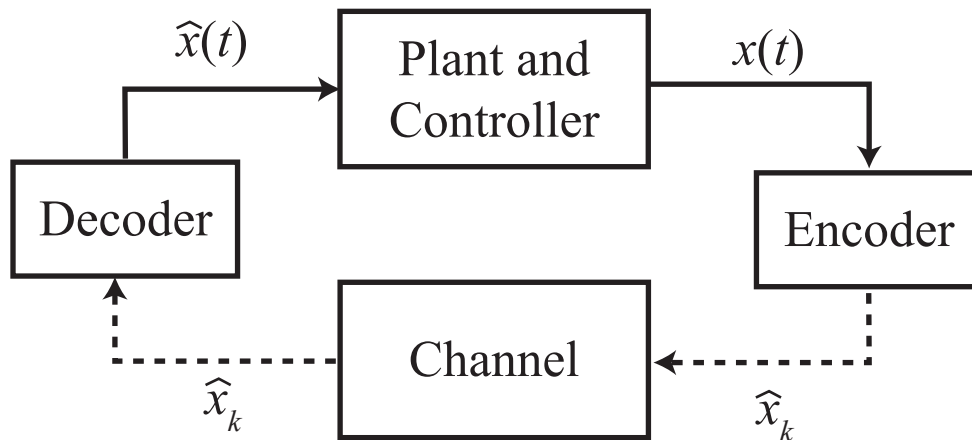

Efficiently Attentive Quantized Event-Triggered Systems

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Sampled Data Control System



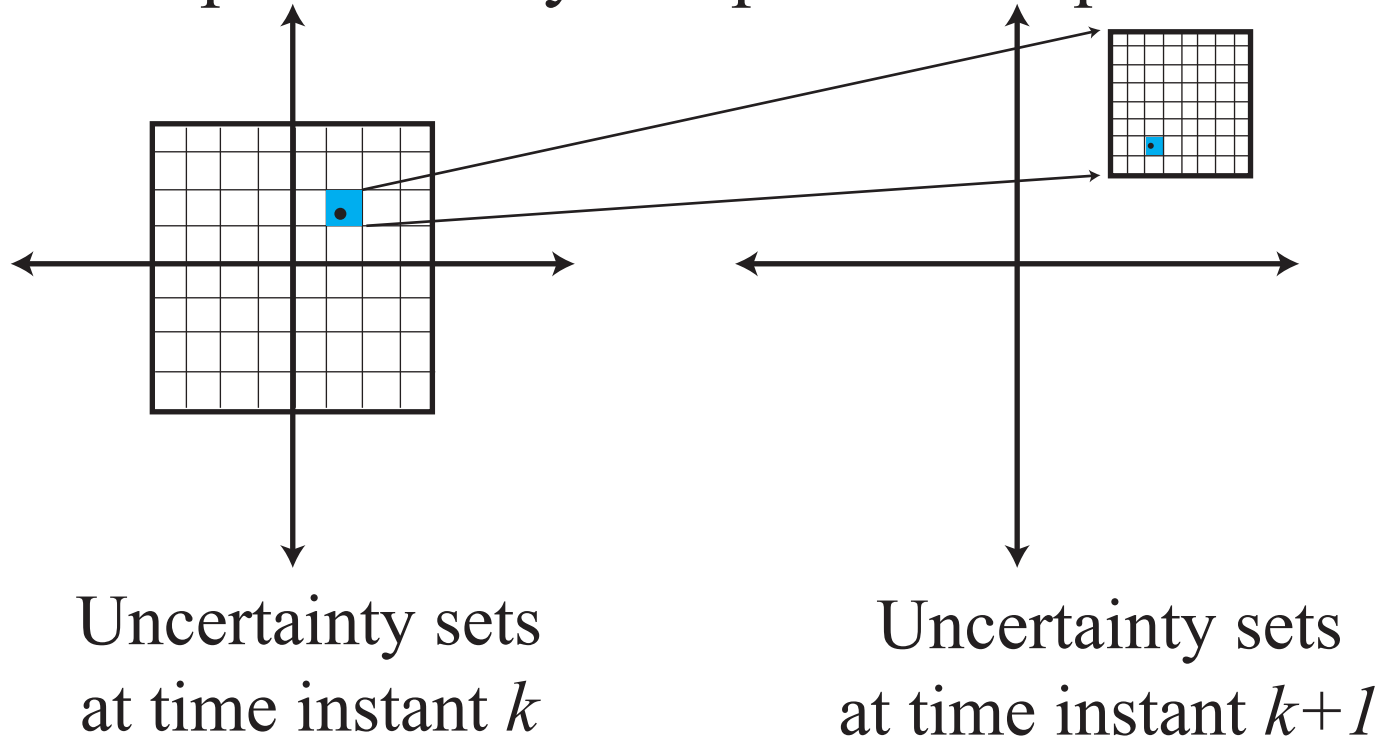
- $\{s_k\}_{k=0}^{\infty}$ = sampling instants
- Plant/Controller state, $x(t)$ satisfies
$$\dot{x}(t) = f(x(t), u(t), w(t)); \quad x(0) = x_0$$
- Sampled and Quantized state, $\hat{x}_k = Q(x(s_k))$
- Reconstructed state $\hat{x}(t) = \hat{x}_k$ for $t \in [s_k, s_{k+1})$.

- Encoder - quantizes and samples system state
- Channel - Lossy with finite bit-rate
- Decoder - zero-order hold
- Plant/Controller - stable under perfect state feedback

Minimum Information for Stabilization?

Dynamically Quantized Feedback

- Discrete-time Linear System with one sample delay
 - state is periodically sampled with quantization^[1,2]



- Necessary and sufficient bit rate for asymptotic stability

$$R \geq \underline{R} = \sum_{i=1}^n \max(0, \log_2 |\lambda_i|)$$

Emulation Method

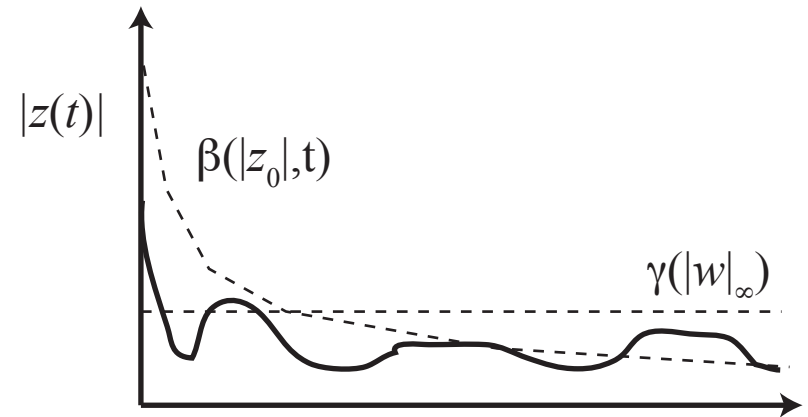
- “Emulation Method” for Sampled System Design
 - design controller, K , that leaves “continuously” sampled system input-to-state stable (ISS)^[3].

There exist functions $\beta \in \mathcal{KL}$ and $\gamma \in \mathcal{K}$ such that

$$|z(t)| < \max\{\beta(|z_0|, t), \gamma(|e|_{\mathcal{L}_\infty})\}$$

$e \in \mathcal{L}_\infty$ and $z(\cdot) : \mathbb{R}^+ \rightarrow \mathbb{R}^n$ satisfies

$$\dot{z}(t) = f(z(t), K(z(t) + e(t)))$$



- Select Sampling Instants to guarantee that the sporadically sampled system is also ISS.

Event-Triggered Feedback^[4]

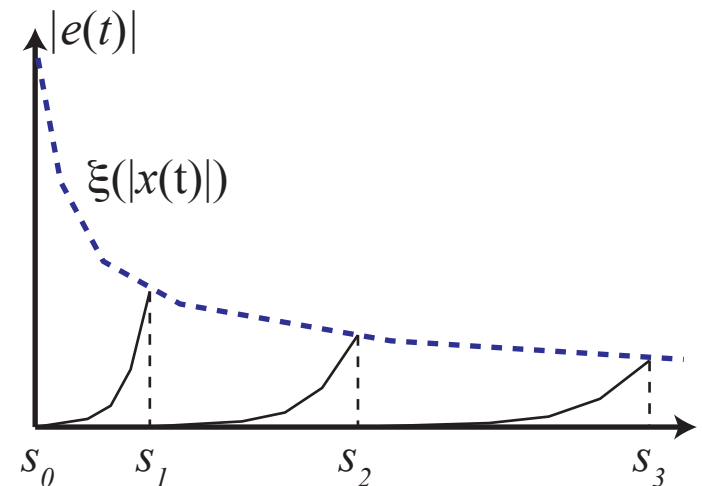
- State is sporadically sampled with no quantization
- Since the "continuously" sampled system is ISS, there exists a positive definite function $V(\cdot) : \mathbb{R}^n \rightarrow \mathbb{R}^+$ and functions $\alpha, \gamma \in \mathcal{K}$ such that

$$\frac{\partial V}{\partial x} f(x, K(x + e)) \leq -\alpha(|x|) + \gamma(|e|)$$

- Select sampling instants, $\{s_k\}_{k=0}^{\infty}$ so that for all time, t ,

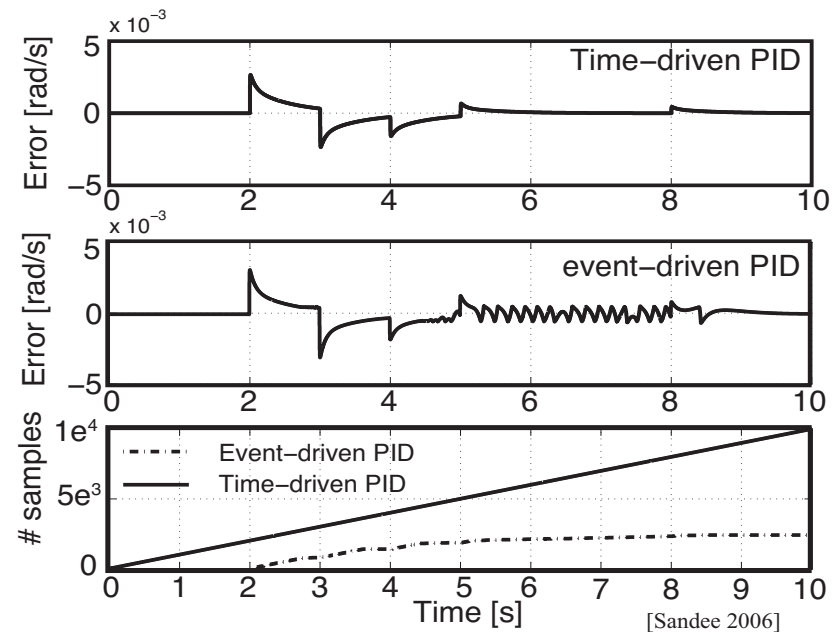
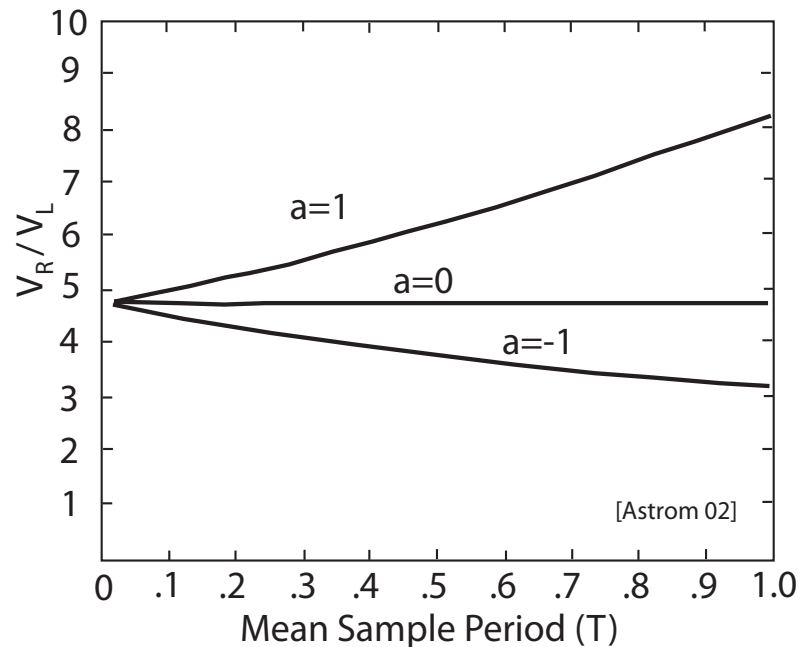
$$|e(t)| \leq \gamma^{-1}((1 - \sigma)\alpha(|x(t)|)) \equiv \xi(|x(t)|)$$

where $0 < \sigma < 1$



Resource Utilization in Event-Triggered Systems

Prior work has suggested that event-triggered systems have lower resource utilization than comparable periodically sampled systems^[5,6,7,8].



But this may not always be the case.

Zeno Sampling and Efficient Attentiveness^[9,10]

- Consider the following system

$$\dot{x}(t) = f(x(t)) + u(t) \text{ with } u(t) = -2f(\hat{x}_k)$$

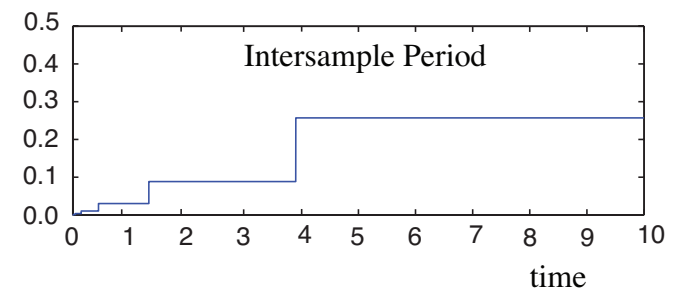
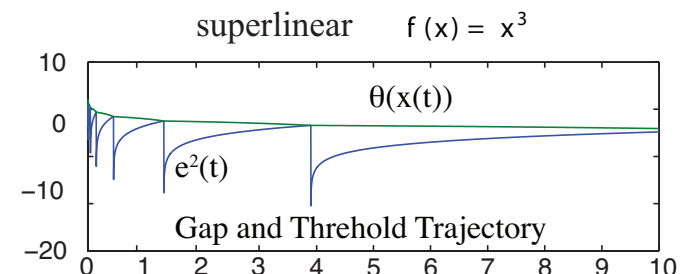
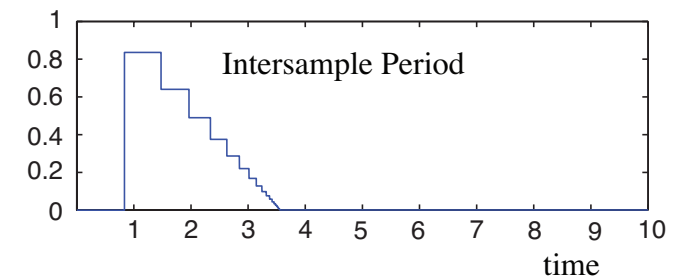
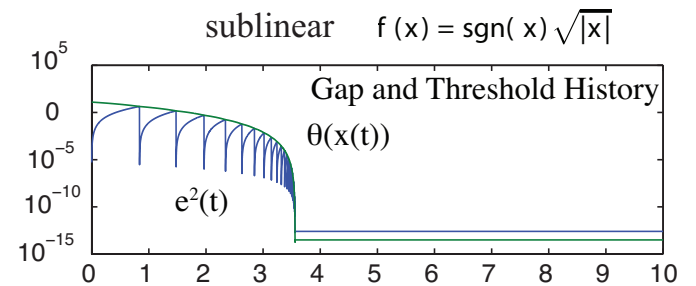
where

sublinear	superlinear
$f(x) = \text{sgn}(x)\sqrt{ x }$	$f(x) = x^3$

- Given sampling instant s_k , the next sampling instant s_{k+1} is the first time when the state leaves

$$\Omega_k = \{x \in \mathbb{R}^n : (\hat{x}_k - x)^2 \leq |x|^2 \equiv \theta(|x|)\}$$

- Zeno-sampling :**
infinite samples over a finite interval
- Efficiently Attentive :**
Intersampling Intervals get larger as system state approach origin.



Dynamic Quantization versus Event-Triggering

- **Differences**

- Single sample delay (Q) versus small delay (ET)
- Utilization measured by bit rate versus inter-sampling time
- Dynamic quantizers achieve minimum stabilizing bit rate, but lower resource utilization not guaranteed for event-triggered systems

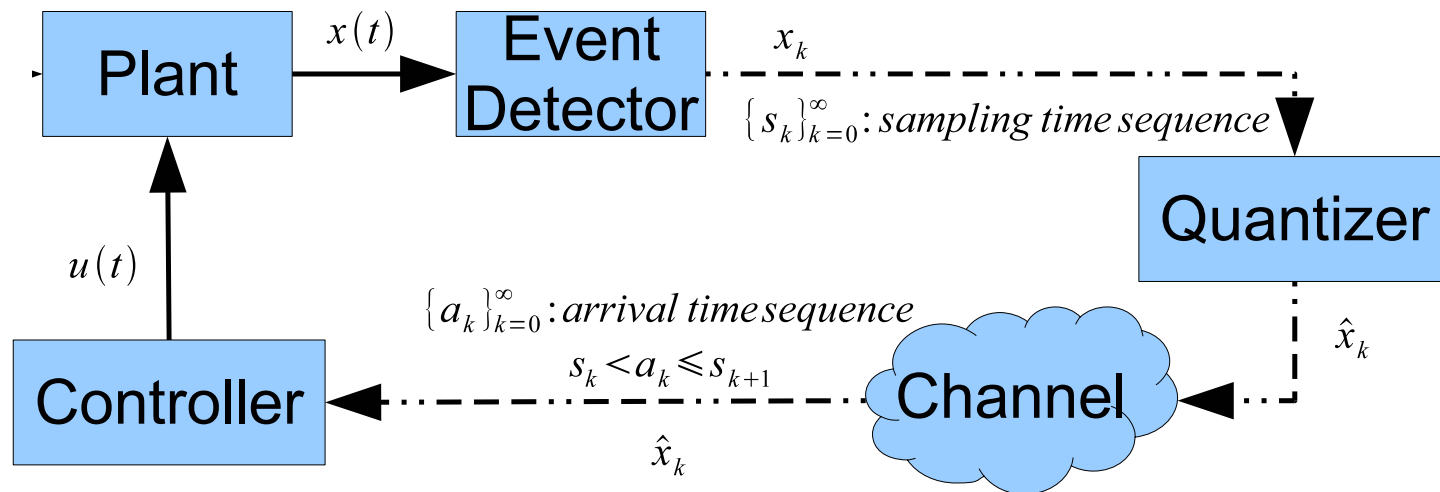
- **Similarities**

- both attempt to reduce “information” over channel.
- quantization in “time” versus “space.
- both discretize a continuous-time control system

- **Objectives:**

- Unification quantization and event-triggering
- Design Event-triggers to control resource utilization by ensuring efficient attentiveness

Quantized Event-Triggered Networked System^[11,12]



- **Sampling Instants**, $\{s_k\}_{k=0}^{\infty}$, and **Arrival Instants** $\{a_k\}_{k=0}^{\infty}$.
- **State Equation** is $\dot{x}(t) = f(x(t), k(\hat{x}_k))$
where \hat{x}_k is the quantized state and $t \in [a_k, a_{k+1})$.
- **Quantized state**, \hat{x}_k , satisfies $|x(s_k) - \hat{x}_k| < E(|x(s_k)|)$
where $E \in \mathcal{K}$ is the **Quantization Error**.
- **Controller**, K , ensures "continuously" sampled system is ISS
with respect to the **gap** $e_k(t) = x(t) - \hat{x}_k$.

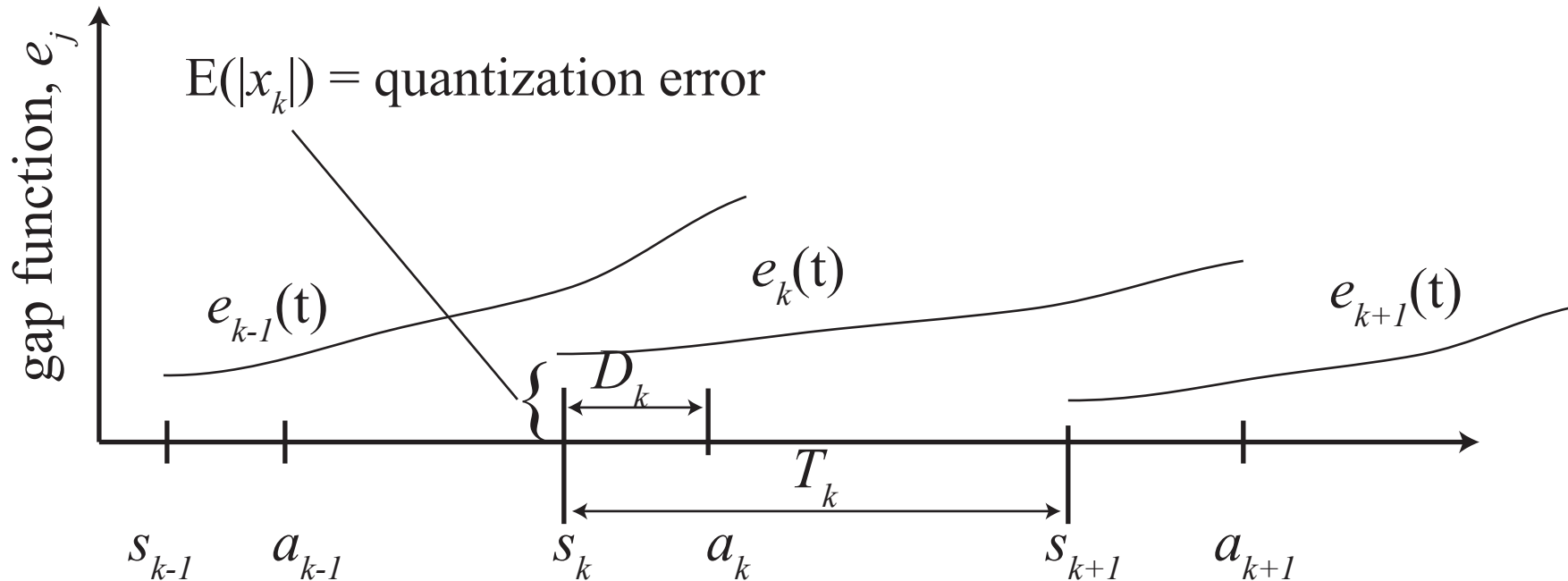
Modeling of Channel Delay and Quantization^[13,14,15]

- Quantization Error, $E(|x_k|)$

$$\text{Delay} = D_k = a_k - s_k$$

$$\text{Inter-sampling Interval} = T_k = s_{k+1} - s_k$$

$$\text{Admissible Sampling: } s_k < a_k < s_{k+1}$$



- What are dynamics of gap function, $e_k(t) = x(t) - \hat{x}_k$?

Dynamics of the Gap Function^[11,12]

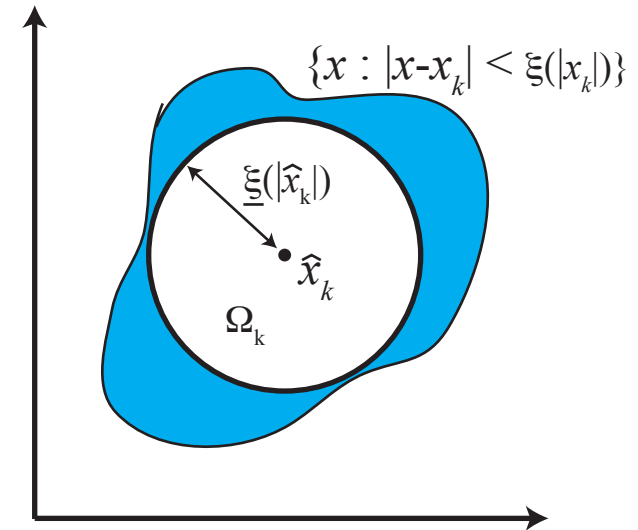
Lipschitz on Compacts: $L_k > 0$ such that

$$|f(x(t), K(\hat{x}_k))| \leq |f(\hat{x}_k, K(\hat{x}_k))| + L_k |e_k(t)|$$

for all $x(t) = \hat{x}_k + e_k(t) \in \Omega_k$ where

$$\Omega_k = \{x \in \mathbb{R}^n : |x| \leq |\hat{x}_k| + \underline{\xi}(|\hat{x}_k|)\}$$

and $\underline{\xi}(s) = \sup \{r : \xi(s-r), s < r\}$

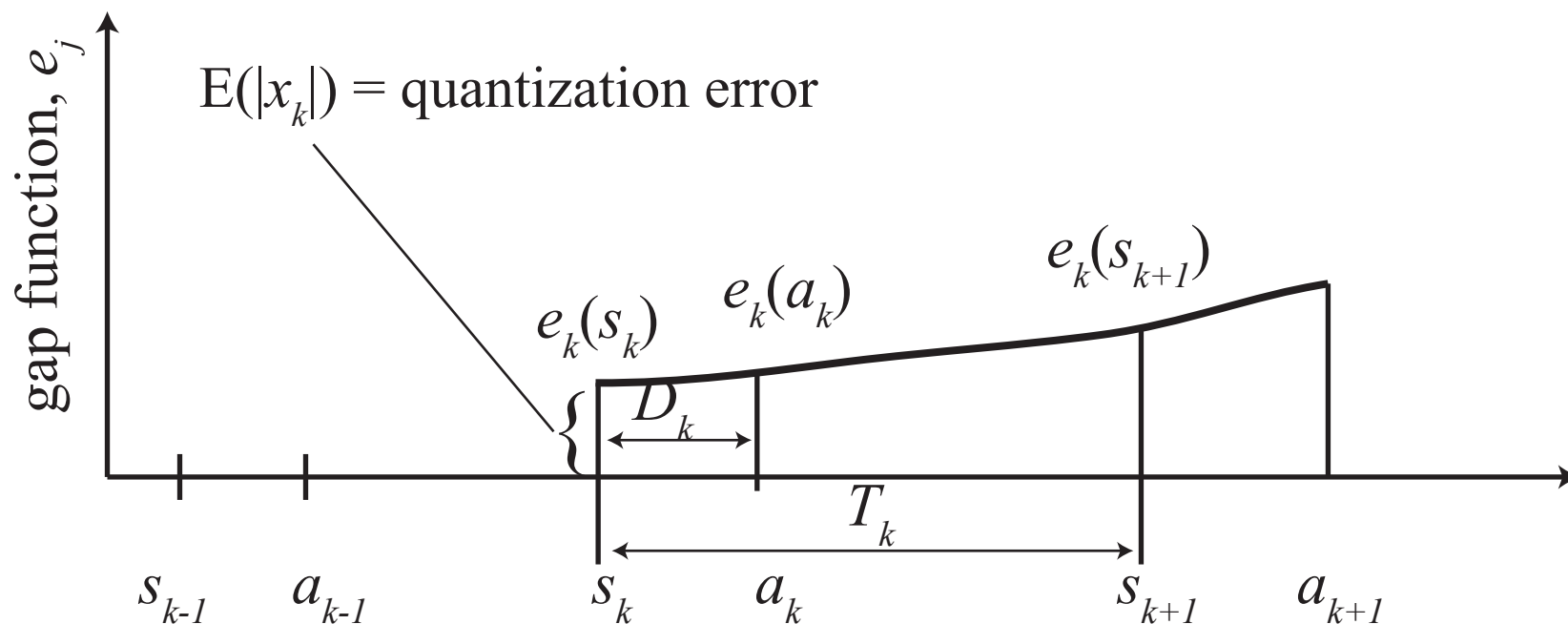


- Since $x(t) \in \Omega_k$, the gap, $e_k(t)$, satisfies the following differential inequality for $t \in [s_k, a_{k+1})$

$$\frac{d|e_k|}{dt} \leq |\dot{e}_k(t)| \leq |f(\hat{x}_k, K(\hat{x}_k))| + L_k |e_k(t)|, \quad e_k(s_k) < E(|x_k|)$$

- This is a linear differential inequality and we can use the comparison principle to bound the gap $|e_k(t)|$

Lower Bound on Inter-sampling Interval^[10,11,13]

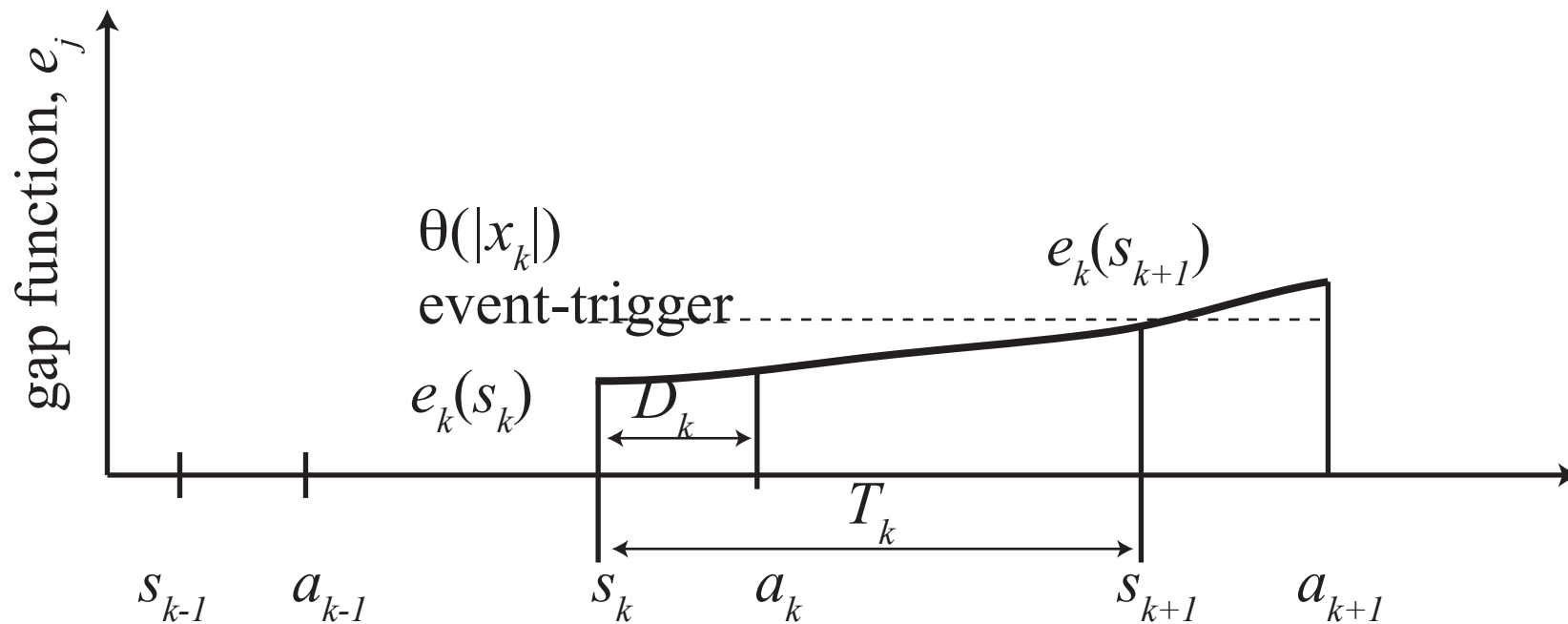


- k th Gap at k th transmission time is $|e_k(s_k)| < E(|x_k|)$
- k th Gap at $k + 1$ st transmission time is

$$|e_k(s_{k+1})| \leq E(|x_k|)e^{L_k T_k} + \frac{\Psi(\hat{x}_k, \hat{x}_{k-1})}{L_k} (e^{L_k T_k} - 1)$$

where $\Psi(\hat{x}, \hat{x}_{k-1}) = |f(\hat{x}_k, K(\hat{x}_k))| + 2|f(\hat{x}_k, K(\hat{x}_{k-1}))|$

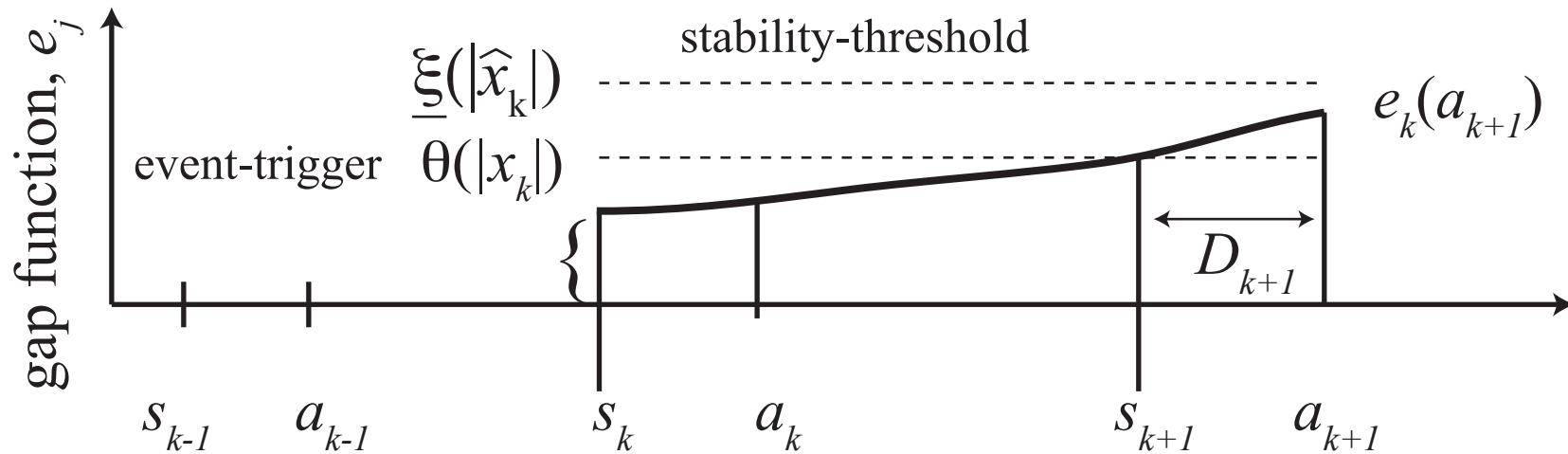
Lower Bound on Inter-sampling Interval



- s_{k+1} , is the first time when $|e_k(t)| = \theta(|\hat{x}_k|)$. We refer to $\theta(\cdot) : \mathbb{R} \rightarrow \mathbb{R}$ as the *event-triggering function*.
- With $|e_k(s_{k+1})| = \theta(|\hat{x}_k|)$, our earlier constraint becomes

$$T_k > \frac{1}{L_k} \left(\ln \left(1 + \frac{L_k \theta(|\hat{x}_k|)}{\Psi(\hat{x}_k, \hat{x}_{k-1})} \right) - \ln \left(1 + \frac{L_k E(|x_k|)}{\Psi(\hat{x}_k, \hat{x}_{k-1})} \right) \right)$$

Upper Bound on Stabilizing Delay^[11,13]



- Using similar techniques, we find

$$|e_k(a_{k+1})| \leq \theta(|\hat{x}_k|) e^{L_k D_{k+1}} + \frac{|f(\hat{x}_k, K(\hat{x}_k))|}{L_k} (e^{L_k D_{k+1}} - 1)$$

- Asymptotic stability requires $|e_k(a_{k+1})| \leq \underline{\xi}(|\hat{x}_k|)$, so the stabilizing delay satisfies

$$D_{k+1} \leq \frac{1}{L_k} \left(\ln \left(1 + L_k \frac{\underline{\xi}(|\hat{x}_k|)}{|f(\hat{x}_k, K(\hat{x}_k))|} \right) - \ln \left(1 + L_k \frac{\theta(|\hat{x}_k|)}{|f(\hat{x}_k, K(\hat{x}_k))|} \right) \right)$$

Asymptotic Stability^[11,12]

Assume that the **event-triggering threshold** θ satisfies

$$\text{Quantization Error} \quad E(s) \leq \theta(s) \leq \underline{\xi}(s) \quad \text{Stability Threshold}$$

for any $s \in \mathbb{R}^+$ and assume the delay $D_k < \min\{\underline{T}_k, \overline{D}_k\}$ where

$$\underline{T}_k = \frac{1}{L_k} \left(\ln \left(1 + \frac{L_k \theta(|\hat{x}_k|)}{\Psi(\hat{x}_k, \hat{x}_{k-1})} \right) - \ln \left(1 + \frac{L_k E(|x_k|)}{\Psi(\hat{x}_k, \hat{x}_{k-1})} \right) \right)$$
$$\overline{D}_k = \frac{1}{L_{k-1}} \left(\ln \left(1 + L_{k-1} \frac{\underline{\xi}(|\hat{x}_{k-1}|)}{|f(\hat{x}_{k-1}, K(\hat{x}_{k-1}))|} \right) - \ln \left(1 + L_{k-1} \frac{\theta(|\hat{x}_{k-1}|)}{|f(\hat{x}_{k-1}, K(\hat{x}_{k-1}))|} \right) \right)$$

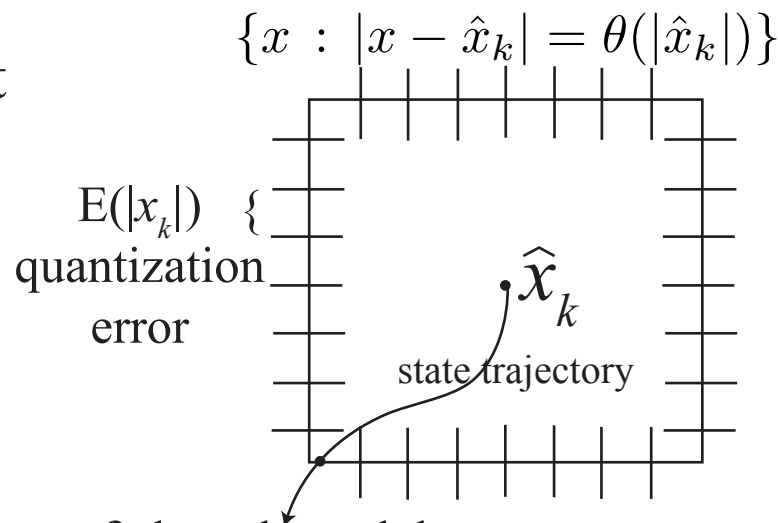
Then the closed-loop quantized event-triggered system is asymptotically stable with an inter-sampling interval T_k which is always bounded below by $\underline{T}_k > 0$.

Stabilizing Bit Rate^[11,12]

- N_k = Number of Bits used to represent sampled state \hat{x}_k

- Assuming $|x|$ is the sup-norm, then

$$N_k = \lceil \log_2 2n \rceil + (n-1) \left\lceil \log_2 \left\lceil \frac{\theta(|\hat{x}_{k-1}|)}{E(|x_k|)} \right\rceil \right\rceil \text{ bits}$$



- To guarantee the asymptotic stability of the closed-loop system, these bits must be delivered within the delay

$$D_k \leq \bar{D}_k \equiv \frac{1}{L_{k-1}} \left(\ln \left(1 + L_{k-1} \frac{\xi(|\hat{x}_{k-1}|)}{|f(\hat{x}_{k-1}, K(\hat{x}_{k-1}))|} \right) - \ln \left(1 + L_{k-1} \frac{\theta(|\hat{x}_{k-1}|)}{|f(\hat{x}_{k-1}, K(\hat{x}_{k-1}))|} \right) \right)$$

- A bit rate stabilizing this system must therefore be

$$R_k = \frac{N_k}{D_k} > \underline{R}_k \equiv \frac{L_k}{\ln 2} (A(|\hat{x}_k|)(n-1) + B(|\hat{x}_k|))$$

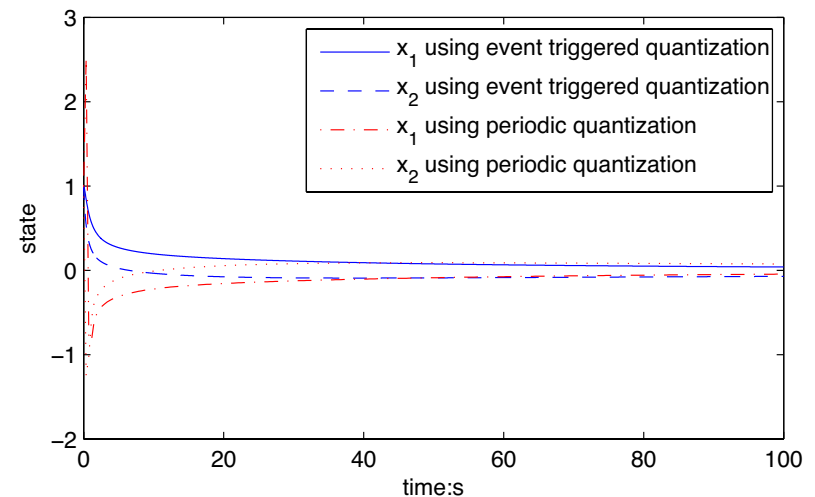
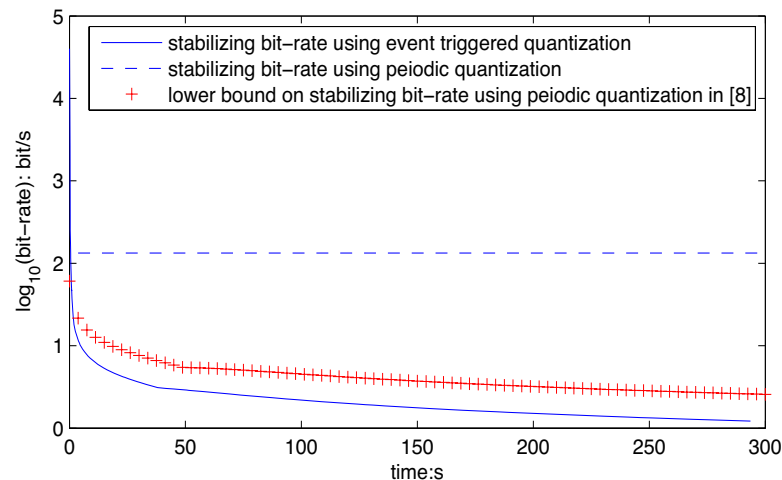
Conditions for Zero Bit Rate^[11]

- In some cases, we can show that the stabilizing bit rate goes to zero as the system approaches its equilibrium point.

- System Equations
$$\begin{aligned}\dot{x}_1 &= x_1^3 + 2x_2^3 + u \\ \dot{x}_2 &= -x_1^3 - x_2^3\end{aligned}$$

Feedback Control
$$u = -2\hat{x}_1^3 - \hat{x}_2^3$$

Switching Condition: $|e_k(t)| = \theta(|\hat{x}_k|) = 0.015|\hat{x}_k|$



Efficiently Attentive Stabilizing Bit Rates^[10,12]

- It can be very difficult to determine the minimal stabilizing bit rate. In such cases, a reasonable option is to require the bit rate to be **efficiently attentive**
- A bit rate is **efficiently attentive** if it is an increasing function of state

Assume that the delay $D_k < \overline{D}_k$, and the event-triggering satisfies

$$E(s) \leq \theta(s) \leq \underline{\xi}(s)$$

Let $\phi_c, \phi_u \in \mathcal{K}$ such that $|f(x, K(x))| \leq \phi_c(|x|)$ and $|K(x)| \leq \phi_u(|x|)$.

If we also know that

$$\lim_{s \rightarrow 0} \frac{\theta(s)}{E(s)} < \infty, \quad \lim_{s \rightarrow 0} \frac{\phi_c(s)}{\theta(s)} < \infty, \quad \lim_{s \rightarrow 0} \frac{\phi_u(s)}{\theta(s)} < \infty$$

then there exists a continuous, positive definite, increasing function $\underline{R}(|\hat{x}_k|)$ such that if the actual bit rate is greater than \underline{R} , then the system is asymptotically stable.

Simulation Example^[12,18]

- This example extends prior work to essentially bounded disturbances

- System Equations

$$\dot{x}_1 = x_1^3 + 2x_2^3 + u_1 + w_1$$

$$\dot{x}_2 = -x_1^3 - x_2^3 + u_2 + w_2$$

Controller

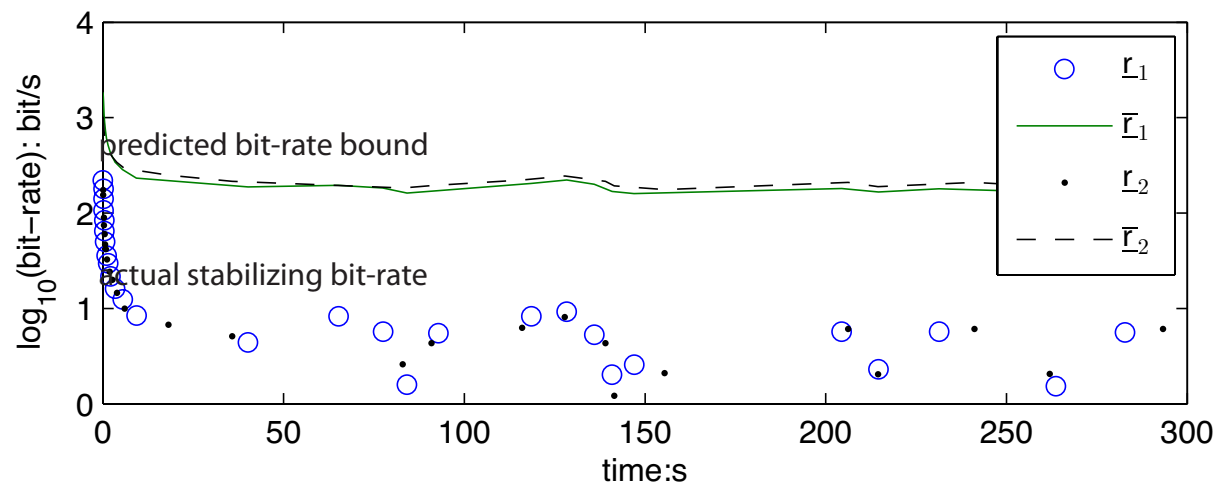
$$u_1 = -3\hat{x}_1^3, \quad u_2 = -3\hat{x}_2^3$$

- Event Trigger and Quantization Map

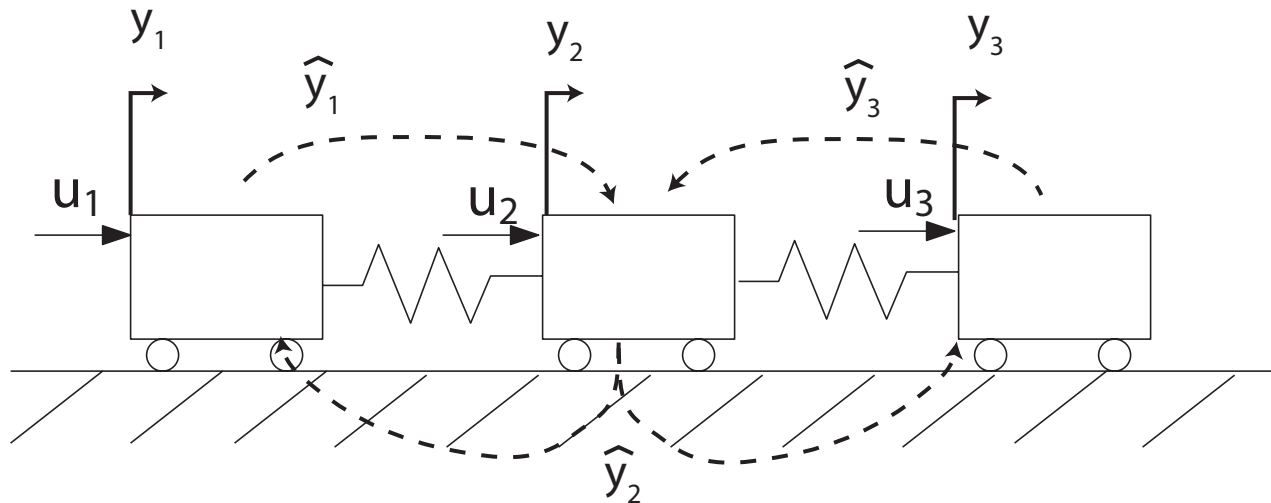
$$\theta_1(s, \bar{w}) = 0.075s^{1.5} + 0.05\bar{w}$$

$$E_1(s, \bar{w}) = 0.025s^{1.5} + 0.017\bar{w}$$

- Predicted bit-rates are conservative bound on actual bit-rates



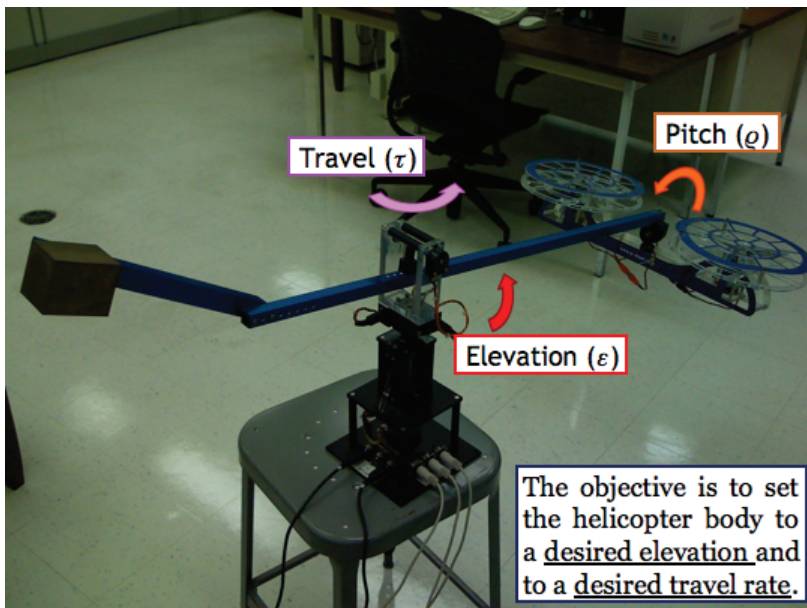
Wireless Networked Control Systems^[15]



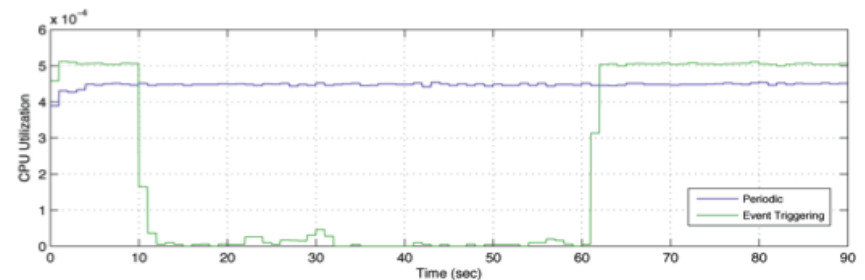
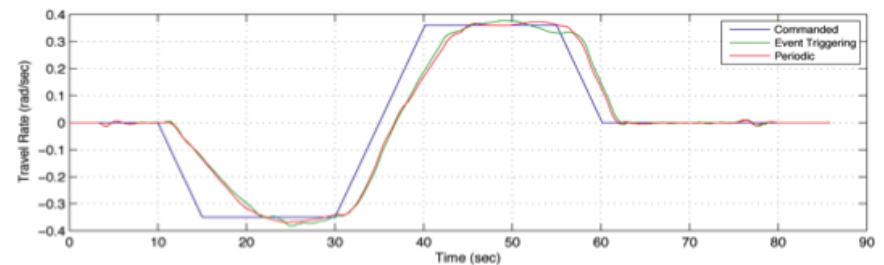
- Event-triggering gives rise to sporadic message streams in wireless networked control systems.
- After an impulsive disturbance is applied to middle cart, one can bound the future inter-sampling times and bit-rate requirements of all controllers.
- Can we use this information to reschedule controller transmission to maintain both overall physical system performance while staying within communication network's capacity limits?

Safety-Critical Systems

- One concern with event-triggered systems is that they are ill suited for safety-critical systems.
- In the presence of disturbances, however, event-triggered solutions must be implemented with a minimum sampling frequency whose size is determined by the “disturbance”.
- With efficiently attentive systems, event frequency increases in the presence of impulsive disturbances.



<http://www.nd.edu/~lemmon/projects/NSF-05-1518/heli-movie/>



Future Directions

- Relax **conservativeness** of bounds^[19].
- Event-trigger design is based on existence of ISS controller (input disturbances). This may not always be possible. One solution may be to extend framework to **iISS controllers**^[16].
- These results provide some guidance on the selection of controllers, quantizers, and event-triggers. We still need to formalize this guidance into a **design procedure**.
- Similarities between dynamic quantization and the quantized event-triggers. When do we achieve the known **necessary and sufficient stabilizing bit rates for linear systems**?
- Efficiently attentive systems provide a basis for co-design of communication/controller in a deterministic setting. Is it possible to extend these ideas to a stochastic setting? One possible approach would involve the use of stochastic ISS concepts^[17].

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