



# ***Deadbeat Response is $\ell_2$ Optimal***

***Vladimír Kučera***

***Czech Technical University in Prague***

***Bill Wolovich Celebratory Event***

***Cancun 2008, Mexico***

# Introduction

A typical linear control strategy in discrete-time systems, *deadbeat control* produces transients that vanish in finite time.

On the other hand, the *linear-quadratic control* stabilizes the system and minimizes the  $l_2$  norm of its transient response.

Quite surprisingly, it is shown that *deadbeat systems are  $l_2$  optimal*, at least for reachable systems.



# Deadbeat Control

Given a linear system  $(A, B)$

$$\mathbf{x}_{k+1} = A\mathbf{x}_k + B\mathbf{u}_k, \quad k = 0, 1, \dots$$

where  $\mathbf{u}_k \in S^m$  and  $\mathbf{x}_k \in S^n$ .

The objective of *deadbeat regulation* is to determine a linear state feedback of the form

$$\mathbf{u}_k = -L\mathbf{x}_k$$

that drives each initial state  $\mathbf{x}_0$  to the origin in a least number of steps.



# Reachability and Controllability

We define the *reachability subspaces* by

$$R_0 = \mathbf{0},$$

$$R_k = \text{range}[B \ AB \ \dots \ A^{k-1}B], \quad k = 1, 2, \dots$$

When  $R_n = S^n$ , the system  $(A, B)$  is said to be *reachable*.

The system  $(A, B)$  is said to be *controllable* if there exists a basis in which

$$A = \begin{bmatrix} A_1 & A_{12} \\ \mathbf{0} & A_2 \end{bmatrix}, \quad B = \begin{bmatrix} B_1 \\ \mathbf{0} \end{bmatrix}$$

where  $(A_1, B_1)$  is reachable and  $A_2$  is nilpotent.



# Reachability Indices

For each  $k = 1, 2, \dots$

let  $S_1, S_2, \dots, S_k$  be a sequence of matrices  
such that

$$\text{range}[BS_1 \quad ABS_2 \quad \dots \quad A^{k-1}BS_k] = \text{range } R_k$$

Therefore  $S_1, S_2, \dots, S_k$  serve to *select a basis* for  $R_k$ .

The *reachability indices*  $r_1, r_2, \dots, r_m$  are defined by

$$r_i = \text{cardinality } \{S_j, j = 1, 2, \dots : \text{rank } S_j \geq i\}$$



# Theorem 1

There exists a deadbeat control law  
if and only if the system  $(A, B)$  is controllable.

Let

$$L_0 = \mathbf{0},$$
$$L_k = L_{k-1} + L'_k (A - BL_{k-1})^k, \quad k = 1, 2, \dots$$

where  $L'_k$  satisfies

$$L'_k [BS_1 \quad ABS_2 \quad \dots \quad A^{k-1}BS_k] = [0 \quad \dots \quad 0 \quad S_k].$$

Then  $L = L_n$  is a deadbeat regulator gain.

The closed-loop system matrix  $A - BL$  is nilpotent.



# Linear Quadratic Regulator

Given a linear system  $(A, B)$

$$\mathbf{x}_{k+1} = A\mathbf{x}_k + B\mathbf{u}_k, \quad k = 0, 1, \dots$$

where  $\mathbf{u}_k \in \mathbb{S}^m$  and  $\mathbf{x}_k \in \mathbb{S}^n$ .

The objective of LQ regulation

is to find a linear state feedback of the form

$$\mathbf{u}_k = -L\mathbf{x}_k$$

that stabilizes the closed-loop system

and, for every initial state  $\mathbf{x}_0$ , minimizes the  $l_2$  norm

of a specified output  $\mathbf{y}_k \in \mathbb{S}^p$  of the form

$$\mathbf{y}_k = C\mathbf{x}_k + D\mathbf{u}_k$$



# Stabilizability and Invertibility

The system  $(A, B)$  is said to be *stabilizable* if there exists a basis in which

$$A = \begin{bmatrix} A_1 & A_{12} \\ \mathbf{0} & A_2 \end{bmatrix}, \quad B = \begin{bmatrix} B_1 \\ \mathbf{0} \end{bmatrix}$$

where  $(A_1, B_1)$  is reachable and  $A_2$  is stable.

The system  $(A, B, C, D)$

$$\mathbf{x}_{k+1} = A\mathbf{x}_k + B\mathbf{u}_k, \quad y_k = C\mathbf{x}_k + D\mathbf{u}_k, \quad k = 0, 1, \dots$$

is said to be *left invertible*

if its transfer function has full column normal rank.





# Invariant Zeros

We further define the *system matrix* as the polynomial matrix

$$S(z) = \begin{bmatrix} zI_n - A & -B \\ C & D \end{bmatrix}$$

and say that a complex number  $\zeta$  is an *invariant zero* of the system  $(A, B, C, D)$  if the rank of  $S(\zeta)$  is strictly less than the normal rank of  $S(z)$ .



## Theorem 2

Suppose that the system  $(A, B)$  is stabilizable.

Suppose that the system  $(A, B, C, D)$  is left invertible and also has no invariant zeros on the unit circle  $|z| = 1$ .

Then, there exists a *unique* LQ regulator gain given by

$$L = (D^T D + B^T X B)^{-1} (B^T X A + D^T C),$$

where  $X$  is the largest nonnegative definite solution of the algebraic Riccati equation

$$X = A^T X A + C^T C - (B^T X A + D^T C)^T (D^T D + B^T X B)^{-1} (B^T X A + D^T C)$$



# Reachability Standard Form

Let the system  $(A, B)$  be reachable,  
with reachability indices  $r_1, r_2, \dots, r_m$ .

Then there exists a similarity transformation  $T$   
that brings the matrices  $A$  and  $B$   
to the *reachability standard form*,

$$A' = TAT^{-1}, \quad B' = TB$$

where  $A'$  is a top-companion matrix  
with nonzero entries in rows  $r_i, i = 1, 2, \dots, m$   
and  $B'$  has nonzero entries  
only in rows  $r_i$  and columns  $j \geq i, i = 1, 2, \dots, m$ .



## **Theorem 3**

Suppose that the system  $(A, B)$  is reachable, with reachability indices  $r_1 \geq r_2 \geq \dots \geq r_m$  and with the matrix  $B$  having rank  $m$ .

Let  $T$  be a similarity transformation that brings  $A$  and  $B$  to the reachability standard form.

Then, the feedback gain  $L$  that is LQ optimal with respect to  $C = T$  and  $D = 0$  is a deadbeat gain.



## ***Proof: Existence***

**We first show that an LQ regulator gain exists that is optimal with respect to  $C = T$  and  $D = 0$ .**

**Indeed, the system  $(A, B)$  is reachable hence stabilizable.**

**The system  $(A, B, T, 0)$  has a transfer function whose normal rank is  $m$ , so it is left invertible.**

**The system matrix**

$$S(z) = \begin{bmatrix} zI_n - A & -B \\ T & 0 \end{bmatrix}$$

**has rank  $n + m$  for all complex numbers  $z$ ,**

**hence  $(A, B, T, 0)$  has no invariant zeros at all.**

**The assumptions of Theorem 2 *are all satisfied*.**



# ***Proof: Polynomial Matrix Fractions***

**Write the transfer function of the system  $(A, B)$  in the polynomial matrix fraction form**

$$(zI_n - A)^{-1} B = Q(z)P^{-1}(z)$$

**For any feedback applied to the system, one obtains**

$$[zI_n - (A - BL)]^{-1} B = Q(z)[P(z) + LQ(z)]^{-1}$$

**The system  $(A, B)$  being reachable, these polynomial matrix fractions are coprime.**

**Thus the matrices  $zI_n - (A - BL)$  and  $P(z) + LQ(z)$  have the *same invariant factors*.**



## ***Proof: Matrix Identity***

Using the polynomial fraction matrices  $P$  and  $Q$ , the algebraic Riccati equation yields the identity

$$\begin{aligned} [P(z^{-1}) + LQ(z^{-1})]^T (D^T D + B^T X B) [P(z) + LQ(z)] \\ = [CQ(z^{-1}) + DP(z^{-1})]^T [CQ(z) + DP(z)] \end{aligned}$$

Define a polynomial matrix  $F$  by the equation

$$F^T(z^{-1})F(z) = [CQ(z^{-1}) + DP(z^{-1})]^T [CQ(z) + DP(z)]$$

in such a way that  $F^{-1}$  is analytic in the domain  $|z| \geq 1$ .

This matrix  $F$  is referred to as the *spectral factor*.



# ***Proof: Reachability Standard Form***

**Bring  $(A, B)$  to the *reachability standard form* using the similarity transformation matrix  $T$ .**

**The corresponding polynomial fraction matrices are related by**

$$P'(z) = P(z), \quad Q'(z) = TQ(z)$$

**and by Structure Theorem,  $Q'$  has the block-diagonal form**

$$Q'(z) = \text{block-diag} \left[ \begin{bmatrix} 1 \\ z \\ \vdots \\ z^{r_1-1} \end{bmatrix}, \begin{bmatrix} 1 \\ z \\ \vdots \\ z^{r_2-1} \end{bmatrix}, \dots, \begin{bmatrix} 1 \\ z \\ \vdots \\ z^{r_m-1} \end{bmatrix} \right].$$





# Proof: Spectral Factorization

The spectral factorization reads

$$\begin{aligned} F^T(z^{-1})F(z) &= Q^T(z^{-1})T^T T Q(z) \\ &= Q'^T(z^{-1})Q'(z) = \text{diag}[r_1, r_2, \dots, r_m] \end{aligned}$$

so that

$$F(z) = \text{diag}[\sqrt{r_1}z^{r_1}, \sqrt{r_2}z^{r_2}, \dots, \sqrt{r_m}z^{r_m}].$$

The matrices  $P(z) + LQ(z)$  and  $zI_n - (A - BL)$

share the same invariant factors  $z^{r_1}, z^{r_2}, \dots, z^{r_m}$ .

Therefore,  $A - BL$  is *nilpotent* with Jordan structure comprising  $m$  nilpotent blocks of sizes  $r_1, r_2, \dots, r_m$ .

This proves that  $L$  is a deadbeat gain.



# Conclusions

- ❖ **Deadbeat control and LQ regulation, two strategies different in nature, are in fact related.**
- ❖ **A deadbeat control law can be obtained by solving a particular LQ regulator problem, at least for reachable systems.**
- ❖ **The LQ optimal regulator gain is unique, whereas the deadbeat feedback gains are not. Only one deadbeat gain is LQ optimal.**
- ❖ **An alternative construction of such a gain is thus available, solving the Riccati equation.**



# References

- C.T. Mullis, “Time optimal discrete regulator gains,”  
*IEEE Trans. Automat. Control*, AC-17, 265–266, 1972.
- A. Saberi, P. Sannuti, and B.M. Chen,  
*H<sub>2</sub> Optimal Control*. London: Prentice-Hall, 1995.
- V. Kučera, “Deadbeat control, pole placement, and LQ  
regulation,” *Kybernetika*, 35, 681–692, 1999.
- V. Kučera, *Analysis and Design of Discrete Linear  
Control Systems*. London: Prentice-Hall, 1991.
- W.A. Wolovich, *Linear Multivariable Systems*.  
New York: Springer, 1974.

